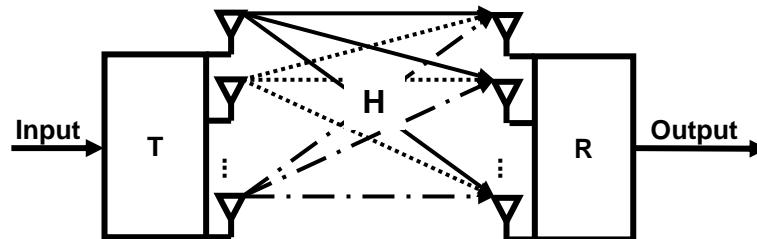

Differential Space-Time Block Codes based on simultaneous phase and amplitude modulation

C. Fellenberg and H. Rohling

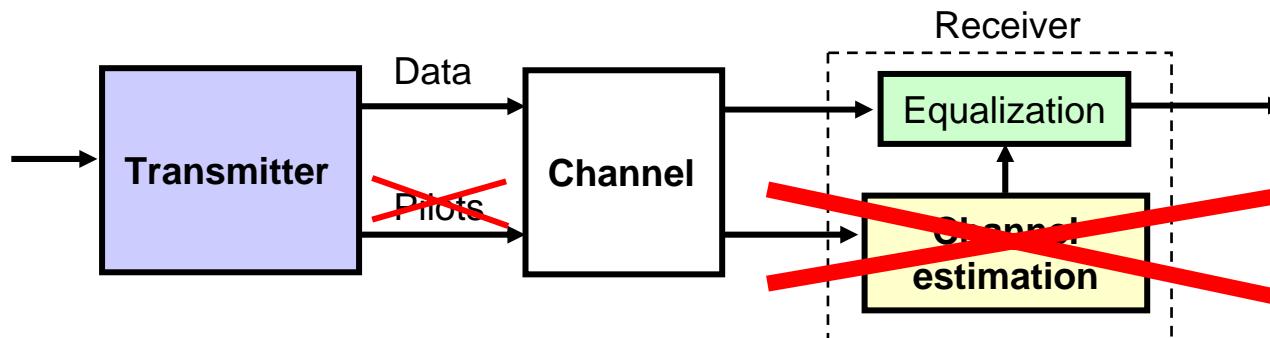
Institute of Telecommunications
Hamburg University of Technology

Motivation for noncoherent transmission systems

MIMO System



Coherent transmission system



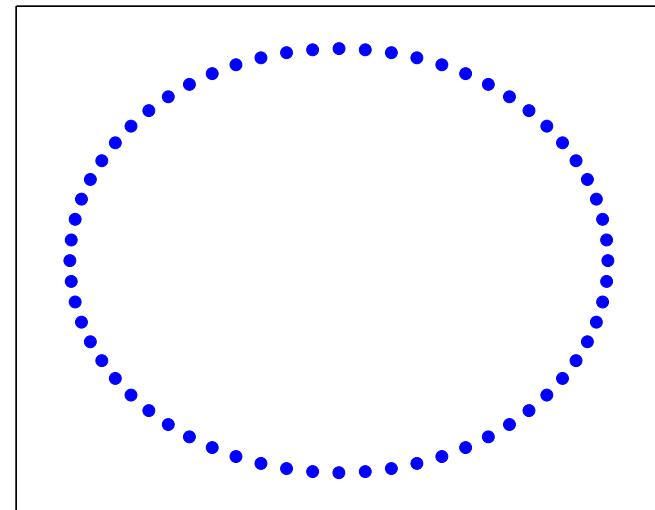
- Drawbacks of coherent transmission systems
 - CSI is never ideal
 - Channel estimation in MIMO is very complex
 - Pilot symbols reduce the useful data rate

Differential Space-Time Block Codes are very attractive in many practical cases

Motivation for higher order modulation

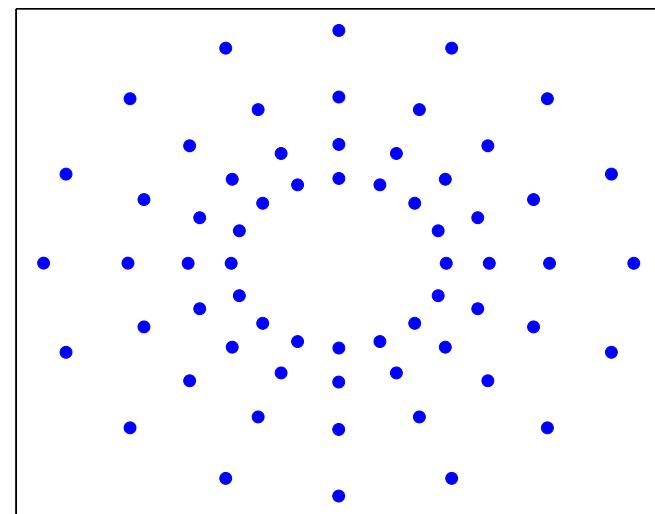
- Common DSTBC: M-PSK modulation
 - small euclidean distance

Example: 64-PSK



- APSK modulation
 - increase euclidean distance

Example: 16-PSK with 4 Amplitudes

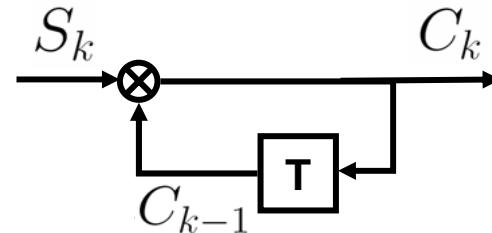


- **Differential Space-Time Block Codes (DSTBC)**
- **Implementation of different DSTBC schemes**
 - **DSTBC based on unitary matrices**
 - M-PSK
 - Differential Amplitude Modulation (DAM)
 - APSK
 - **DSTBC based on non-unitary matrices**
 - 4A16PSK
- **Simulation results**
- **Conclusion**

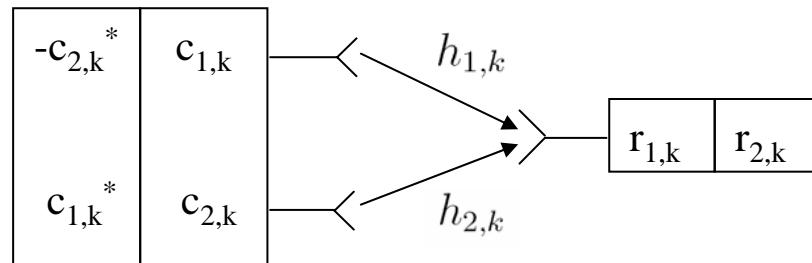
- Symbol matrix S_k

$$S_k = \begin{bmatrix} s_{1,k} & s_{2,k} \\ -s_{2,k}^* & s_{1,k}^* \end{bmatrix}$$

DSTBC: $C_k = S_k C_{k-1}$



- System model



$$\begin{pmatrix} r_{1,k} & -r_{2,k}^* \\ r_{2,k} & r_{1,k}^* \end{pmatrix} = C_k \begin{pmatrix} h_{1,k} & -h_{2,k}^* \\ h_{2,k} & h_{1,k}^* \end{pmatrix} + \begin{pmatrix} w_{1,k} & -w_{2,k}^* \\ w_{2,k} & w_{1,k}^* \end{pmatrix}$$

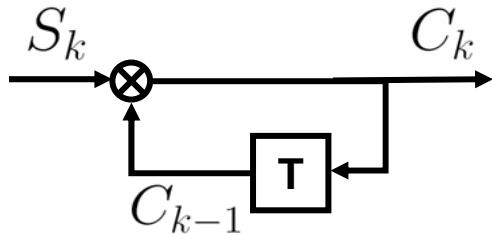
Matrix notation

$R_k = C_k H_k + W_k$

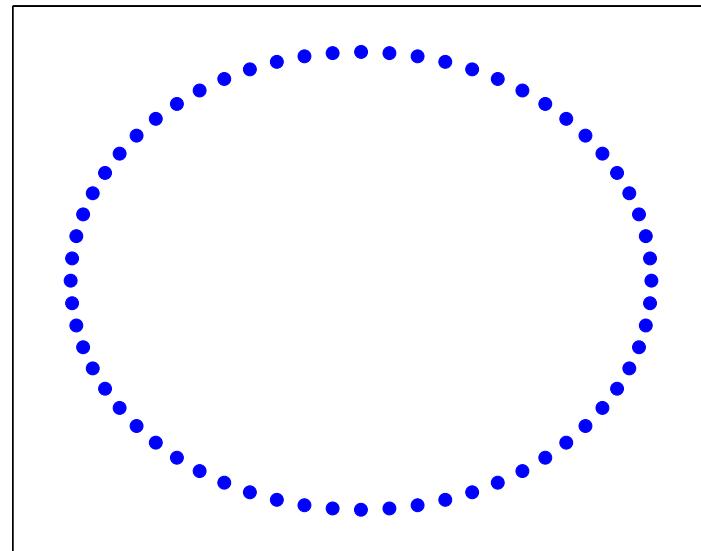
Unitary DSTBC based on 64-PSK

- Coding: Rate=6bits/s/Hz

$$C_k = S_k C_{k-1}$$



Constellation diagram for S_k



- Decoding:

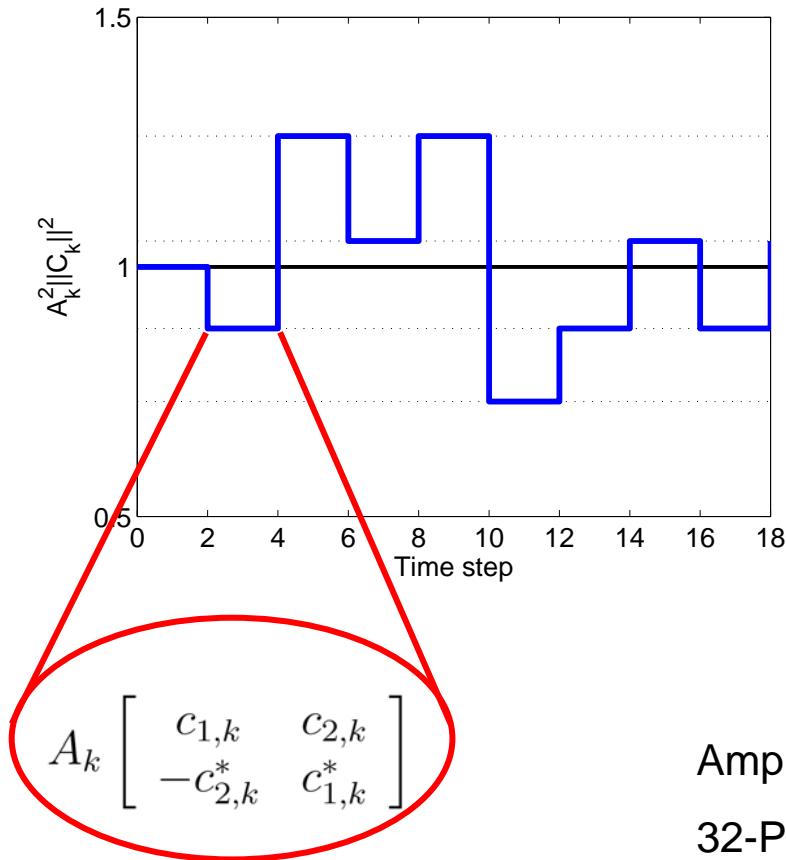
$$R_k R_{k-1}^*$$

64-PSK: 2×6 bits/Symbol matrix

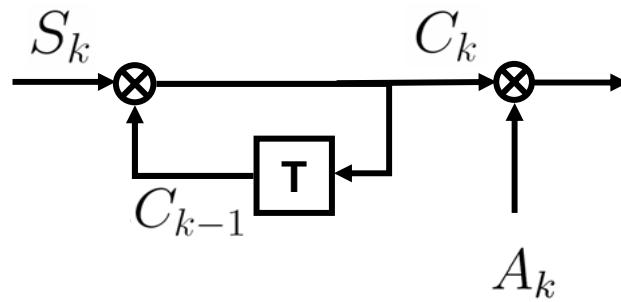
$$R_k R_{k-1}^* = C_k H H^* C_{k-1}^* + noise = \|H\|^2 S_k + noise$$

Unitary DSTBC based on 4-DAM + 32-PSK

- 4-Differential Amplitude Modulation (4-DAM)



$$C_k = S_k C_{k-1}$$



Amplitude: 2 bits/Symbol matrix
32-PSK: 2*5 bits/Symbol matrix

Decoding

1. Decode amplitude

$$\frac{\|R_k\|}{\|R_{k-1}\|} \simeq \frac{\|H\|\|C_k\|A_k}{\|H\|\|C_{k-1}\|A_{k-1}} = \frac{A_k}{A_{k-1}}$$

2. Decode phase $R_k R_{k-1}^*$

$$\begin{aligned} R_k R_{k-1}^* &= A_k C_k H H^* C_{k-1}^* A_{k-1} + noise \\ &= A_k A_{k-1} \|H\|^2 S_k + noise \end{aligned}$$

Unitary DSTBC based on APSK

- Unitary matrices used in common DSTBC

$$S_k = \frac{1}{\sqrt{2}} \begin{bmatrix} e^{j\phi_{1,k}} & e^{j\phi_{2,k}} \\ -e^{-j\phi_{2,k}} & e^{-j\phi_{1,k}} \end{bmatrix} \Rightarrow S_k^* S_k = I$$

- More general unitary matrices

$$S_k = \begin{bmatrix} A_{1,k} e^{j\phi_{1,k}} & A_{2,k} e^{j\phi_{2,k}} \\ -A_{2,k} e^{-j\phi_{2,k}} & A_{1,k} e^{-j\phi_{1,k}} \end{bmatrix} \Rightarrow S_k^* S_k = (A_{1,k}^2 + A_{2,k}^2)I$$

A_{1,k} and A_{2,k} can be used to code information as long as A_{1,k}² + A_{2,k}² = 1 is fulfilled

Unitary DSTBC based on APSK

- Coding: Rate=6bits/s/Hz

$$C_k = S_k C_{k-1}$$

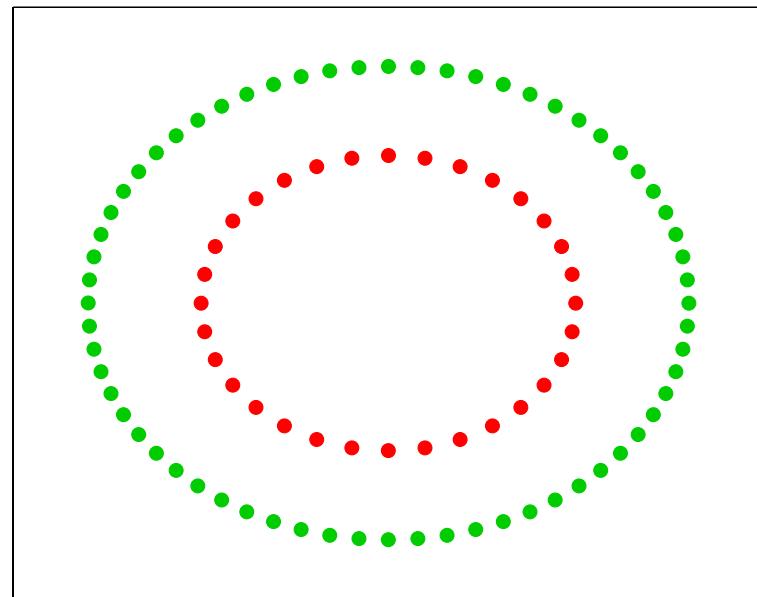
Input bit	$S_{1,k}$	$S_{2,k}$
0	32-PSK	64-PSK
1	64-PSK	32-PSK

$$|s_{1,k}|^2 + |s_{2,k}|^2 = 1$$

Amplitude: 1 bit/Symbol matrix

M-PSK: 5+6 bits/Symbol matrix

Constellation diagram for S_k



- Decoding $R_k R_{k-1}^* = C_k H H^* C_{k-1}^* + noise = \|H\|^2 S_k + noise$

	$S_{1,k}$	$S_{2,k}$	Output bit
$ S_{1,k} < S_{2,k} $	32-PSK	64-PSK	0
$ S_{1,k} > S_{2,k} $	64-PSK	32-PSK	1

Non-unitary DSTBC based on 4A16PSK

- Coding: Rate=6 bits/s/Hz

$$C_k = S_k C_{k-1}$$

$$S_k S_k^* = (|s_{1,k}|^2 + |s_{2,k}|^2) I$$



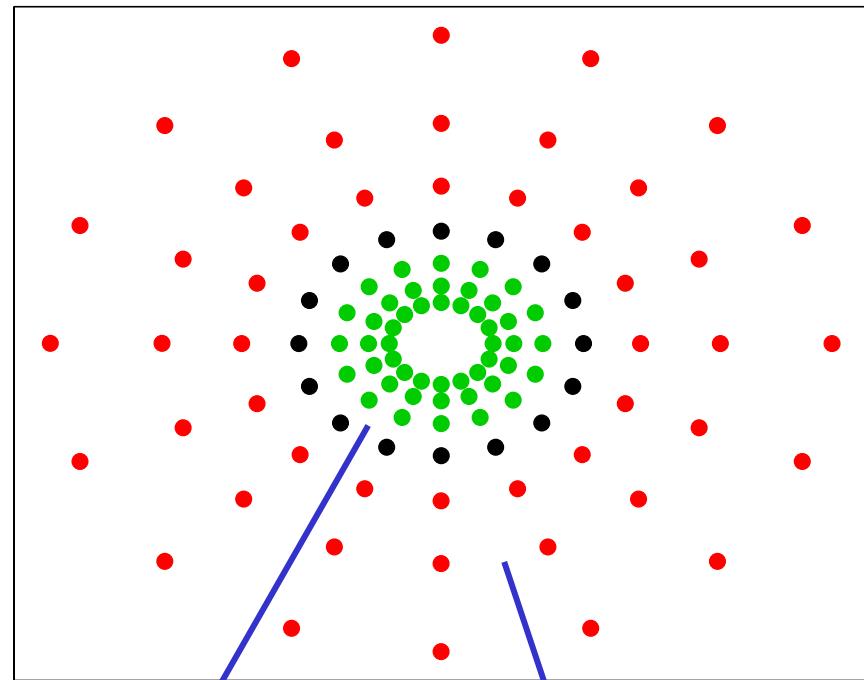
Envelope Control of C_k necessary!!!

Approach: Constellation Expansion

Amplitude: 2*2 bit/Symbol matrix

16-PSK: 2*4 bits/Symbol matrix

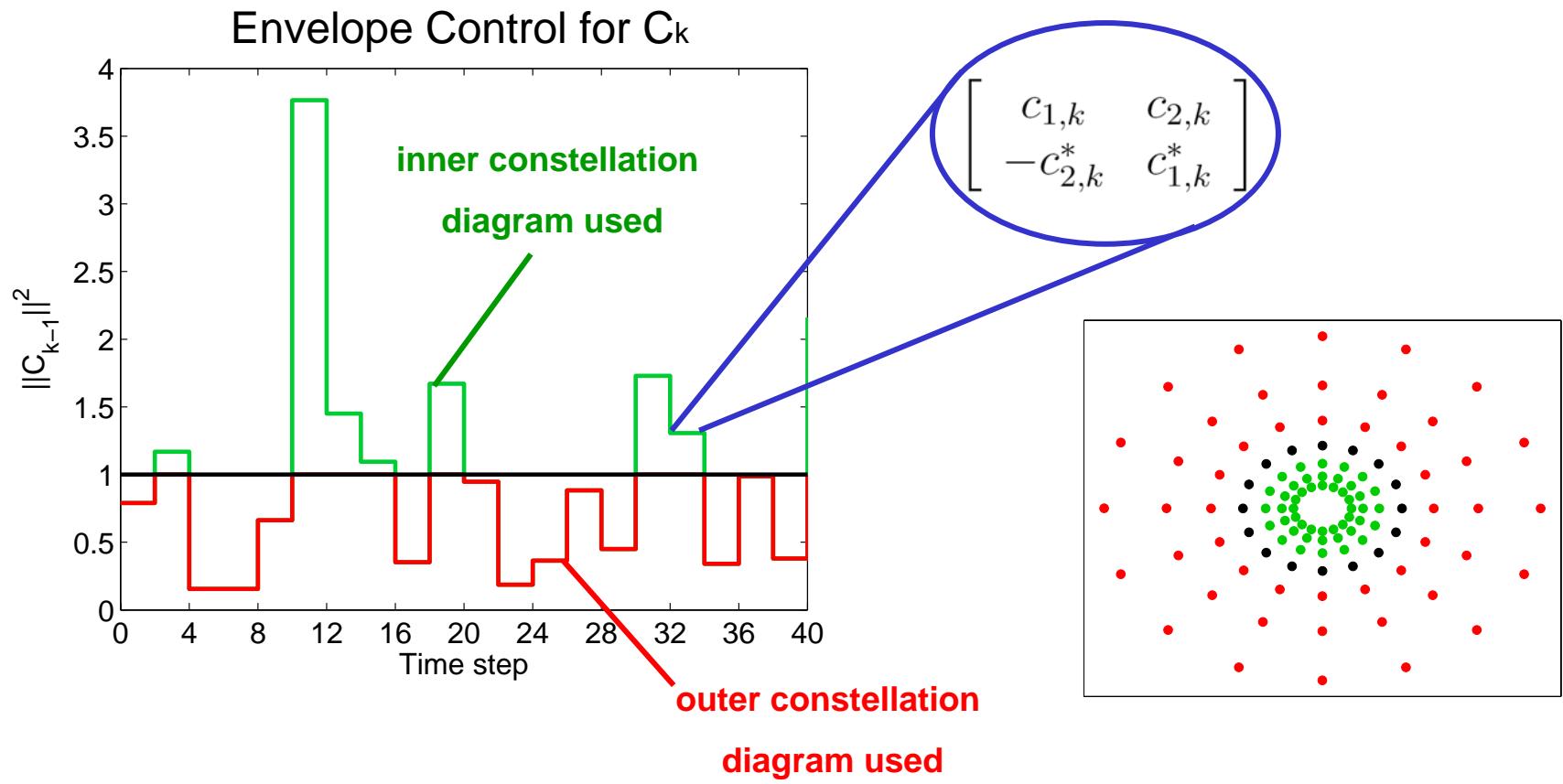
Constellation diagram for S_k



inner constellation diagram

outer constellation diagram

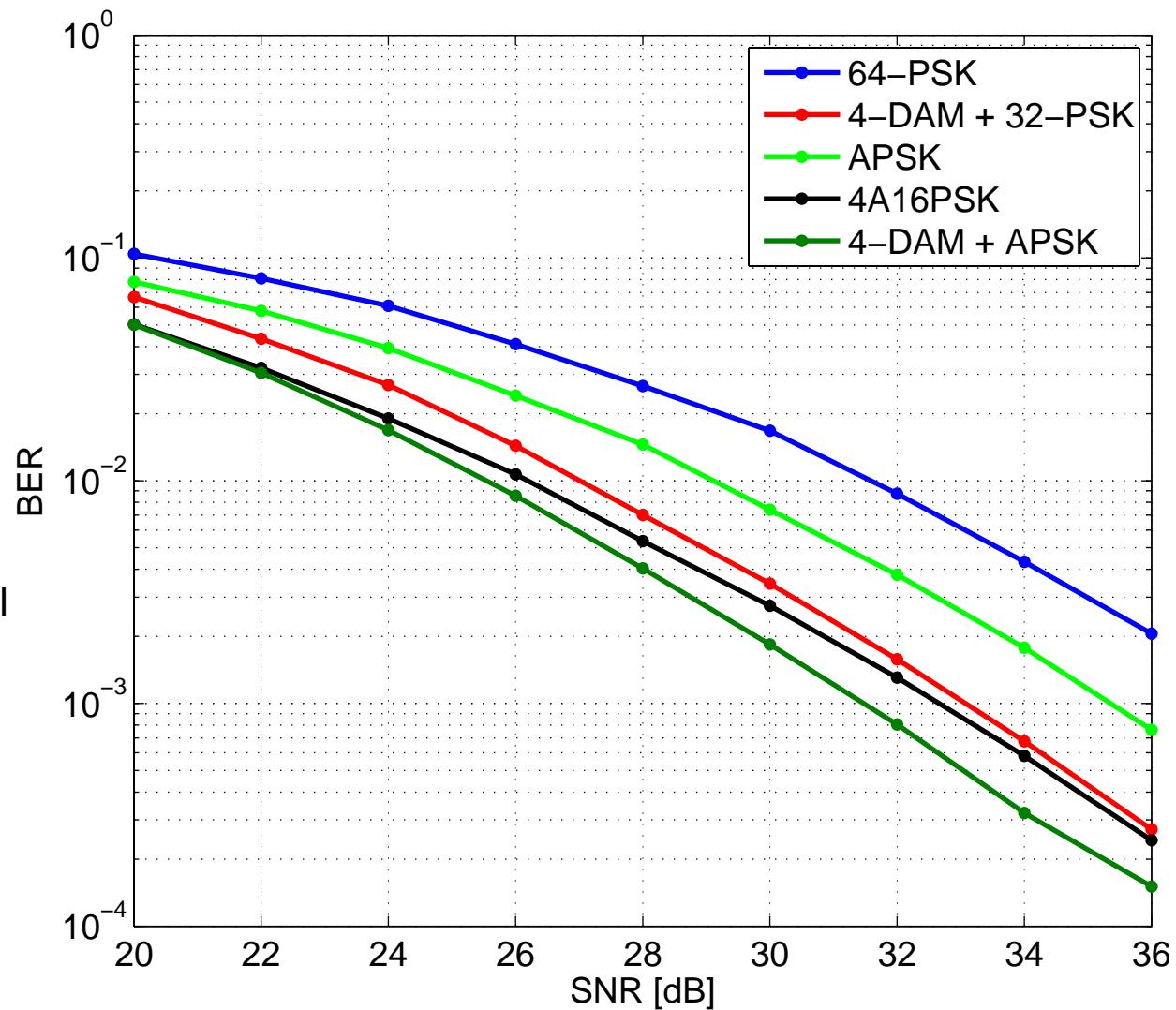
Non-unitary DSTBC based on 4A16PSK



- New decoding required

$$R_k R_{k-1}^{-1} = C_k H H^{-1} C_{k-1}^{-1} + \text{noise} = S_k + \text{noise}$$

Simulation results



Parameters:

- 2 Transmit antennas
- 1 Receive antenna
- Rate: 6bits/s/Hz
- uncoded transmission
- Rayleigh fading channel

Conclusion

- Different modulation alternatives for DSTBC have been considered
- The performance of different modulation alternatives were analyzed and compared for independent Rayleigh fading channel.
- For high bandwidth efficiency of 6 bits/s/Hz the combination 4-DAM + APSK is found to be the best alternative